

General Description

The LH0044 Series is an excellent choice for a wide range of precision applications including strain gauge bridges, thermocouple amplifiers, and ultrastable reference ampli-

Features

- Low input offset voltage 25 μ V max
- Excellent long-term stability $\pm 1\mu$ V/month max
- Low offset drift 0.5 μ V/ $^{\circ}$ C max
- Very low noise 0.7 μ Vp-p max 0.1 Hz to 10 Hz
- High CMRR and PSRR 120 dB min
- High open loop gain 120 dB min
- Wide common-mode range ± 13 V min
- Wide supply voltage range ± 2 V to ± 20 V

The circuit diagram shows a precision rectifier using an operational amplifier (A₁) and a differential pair (Q₁, Q₂). The op-amp's non-inverting input (+) is connected to a voltage divider consisting of resistors R₁ (50k) and R₂ (50k) connected to the 'OVER COMP' supply (pin 1), and a feedback path from the output (pin 6) through a capacitor C₁ (100 pF). The op-amp's inverting input (-) is connected to the output of the differential pair. The differential pair consists of two NPN transistors, Q₁ and Q₂, whose emitters are connected to a common emitter resistor R₇ (50k) leading to ground. The bases of Q₁ and Q₂ are connected to the inputs of the precision rectifier. The 'INVERTING INPUT' (pin 2) is connected to the base of Q₁ through resistor R₅ (500). The 'NON-INVERTING INPUT' (pin 3) is connected to the base of Q₂ through resistor R₆ (500). Diodes D₁ and D₂ are connected in series between the inputs and the bases of Q₁ and Q₂ respectively. The output of the differential pair is taken from the collector of Q₂ (pin 6). The circuit is powered by a 'V⁺' supply (pin 7) and a 'V⁻' supply (pin 4). A current source I₁ is connected between the 'OVER COMP' supply and the collector of Q₆, which is also connected to the base of Q₅. A diode D₃ (1.3V) is connected between the collector of Q₅ and the 'V⁻' supply. A resistor R₃ (100) is connected between the collector of Q₁ and the 'V⁻' supply, and a resistor R₄ (100) is connected between the collector of Q₂ and the 'V⁻' supply. A 'COMP' pin (pin 8) is also shown.

TL/K/5551-1

Absolute Maximum Ratings

If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/Distributors for availability and specifications. (Note 6)

Supply Voltage	± 20V
Power Dissipation	600 mW
Differential Input Voltage (Note 4)	± 1V
Input Voltage (Note 5)	± 15V

Output Short-Circuit Duration

Continuous

Operating Temperature Range

LH0044 -55°C to +125°C

LH0044AC, LH0044B, LH0044C -25°C to +85°C

Storage Temperature Range

-65°C to +150°C

Lead Temperature (Soldering, 10 sec.)

260°C

ESD rating to be determined.

DC Electrical Characteristics (Note 1)

Parameter	Conditions	Limits						Units
		LH0044AC			LH0044/ LH0044B/LH0044C			
		Min	Typ	Max	Min	Typ	Max	
Input Offset Voltage	T _A = 25°C, R _S = 50Ω, V _{CM} = 0V LH0044C Only		8	25		12	50 100	μV μV
Input Offset Voltage	R _S = 50Ω, V _{CM} = 0V LH0044A and LH0044B Only			55 75			180 80	μV μV
Average Input Offset Voltage Drift	T _{Min} ≤ T _A ≤ T _{Max} LH0044B Only		0.1	0.5		0.2	1.3 0.5	μV/°C μV/°C
Long-Term Stability	(Note 2)		0.2	1		0.3	2	μV/month
Input Noise Voltage (Note 3)	BW = 0.1 Hz to 10 Hz, R _S = 50Ω R _S = 10 kΩ Imbalance		0.35 0.50	0.7 0.9		0.35 0.50	0.8 1.0	μVp-p μVp-p
Thermal Feedback Coefficient			0.005			0.005		μV/mW
Open Loop Voltage Gain	R _L = 10 kΩ	120	145		114	140		dB
Common-Mode Rejection Ratio	−10V ≤ V _{CM} ≤ +10V	120	145		114	140		dB
Power Supply Rejection Ratio	±3V ≤ V _S ≤ ±18V	120	145		114	140		dB
Input Voltage Range		±13	±13.8		±12	±13.5		V
Output Voltage Swing	R _L = 10 kΩ	±13	±13.7		±12	±13.5		V
Input Offset Current	25°C ≤ T _A ≤ T _{Max} T _{Min} ≤ T _A < 25°C		1.0	2.5 5.0		1.5	5.0 10.0	nA nA
Average Input Offset Current Drift			5	40		15	80	pA/°C
Input Bias Current	25°C ≤ T _A ≤ T _{Max} T _{Min} ≤ T _A < 25°C		8.5	15 50		10	30 100	nA nA
Average Input Bias Current Drift			50	300		100	600	pA/°C
Differential Input Impedance		5	10		2.5	8		MΩ
Common Mode Input Impedance			2 x 10 ¹¹			2 x 10 ¹¹		Ω
Supply Current	I _L = 0		0.9	3.0		1.0	4.0	mA
Power Dissipation			27	90		30	120	mW

AC Electrical Characteristics $T_A = 25^\circ\text{C}$, $V_S = \pm 15\text{V}$

Parameter	Conditions	Typ	Units
Input Noise Voltage	$R_S = 1\text{ k}\Omega$, $f_O = 10\text{ Hz}$	11	$\text{nV}/\sqrt{\text{Hz}}$
	$R_S = 1\text{ k}\Omega$, $f_O = 1\text{ kHz}$	9	$\text{nV}/\sqrt{\text{Hz}}$
Slew Rate	$A_V = +1$, $R_L = 10\text{ k}\Omega$, $V_{IN} = \pm 10\text{V}$	0.06	$\text{V}/\mu\text{s}$
Large Signal Bandwidth	$A_V = +1$, $R_L = 10\text{ k}\Omega$, $V_{IN} = \pm 10\text{V}$	1	kHz
Overload Recovery Time	$A_V = +100$, $V_{IN} = -100\text{ mV}$, $\Delta V_{IN} = 200\text{ mV}$	5	μs
Small Signal Bandwidth	$A_V = +1$, $R_L = 10\text{ k}\Omega$	400	kHz
Small Signal Rise Time	$A_V = +1$, $R_L = 10\text{ k}\Omega$, $V_{IN} = 10\text{ mV}$	2.5	μs
Overshoot	$A_V = +1$, $R_L = 10\text{ k}\Omega$, $V_{IN} = 10\text{ mV}$, $C_L = 100\text{ pF}$	10	%

Note 1: All specifications apply for all device grades, at $V_S = \pm 15\text{V}$, and from T_{Min} to T_{Max} unless otherwise specified. T_{Min} is -55°C and T_{Max} is $+125^\circ\text{C}$ for the LH0044. T_{Min} is -25°C and T_{Max} is $+85^\circ\text{C}$ for the LH0044AC, LH0044B and LH0044C. Typicals are given for $T_A = 25^\circ\text{C}$.

Note 2: This parameter is not 100% tested; however, 90% of the devices are guaranteed to meet this specification after one month of operation and after initial turn-on stabilization.

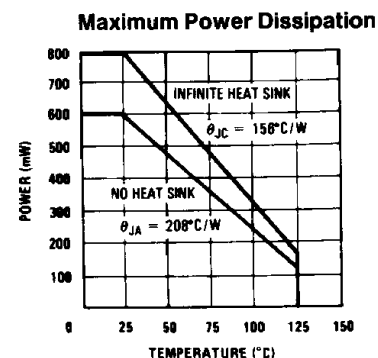
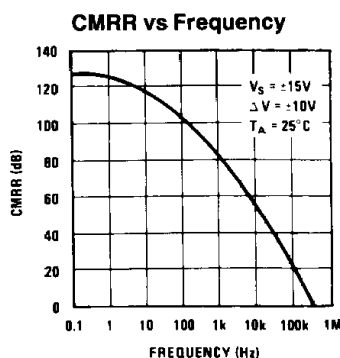
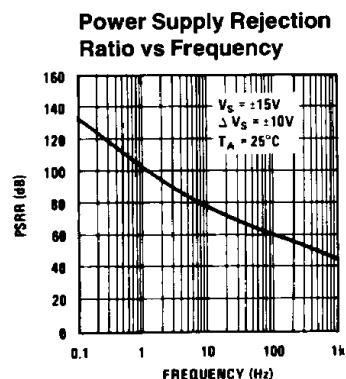
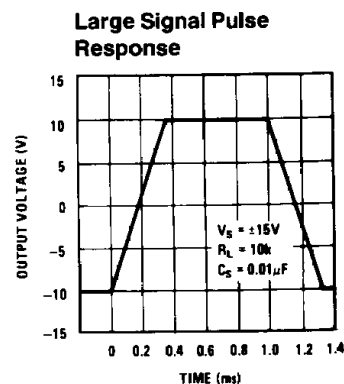
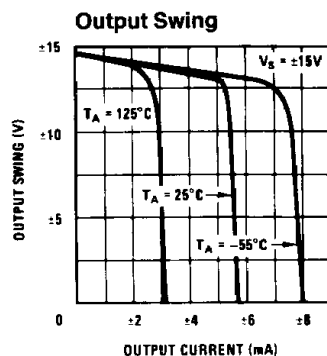
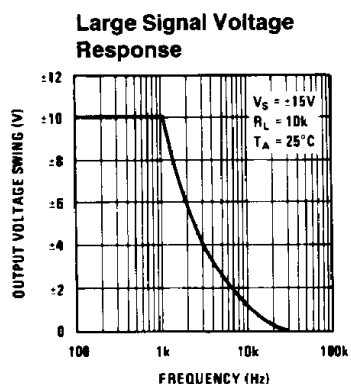
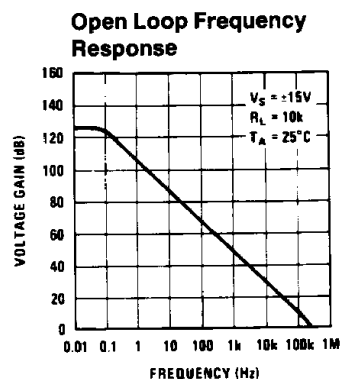
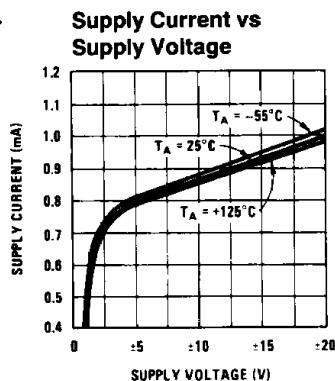
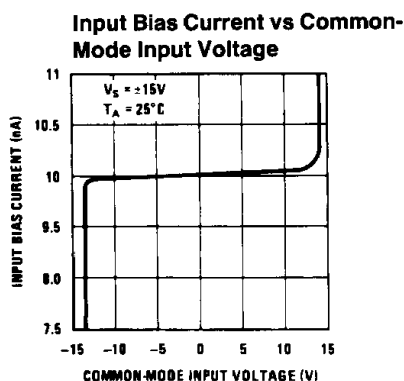
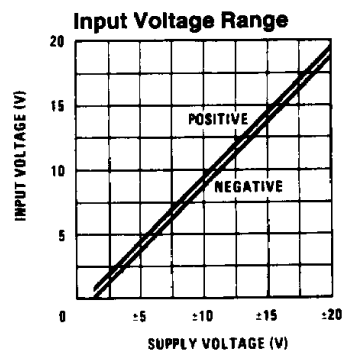
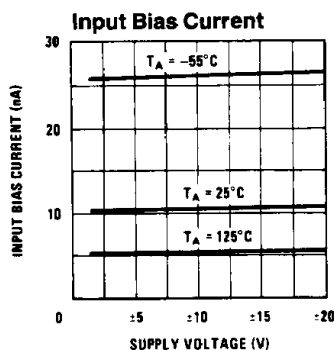
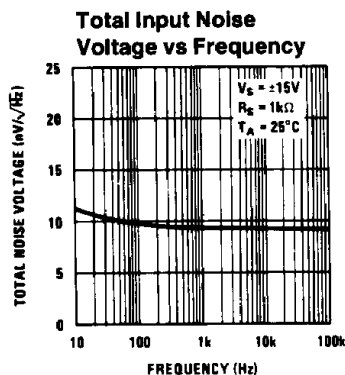
Note 3: Noise is 100% tested on the LH0044AC and LH0044B only. 90% of the LH0044 and LH0044C devices are guaranteed to meet this specification.

Note 4: The inputs are shunted by back-to-back diodes for over-voltage protection. Excessive current will flow for differential input voltages in excess of 1V. Input current should be limited to less than 1 mA.

Note 5: For supply voltages less than $\pm 15\text{V}$, the absolute maximum input voltage is equal to the supply voltage.

Note 6: Refer to RETS0044AH for LH0044AH military specifications and RETS0044H for LH0044H military specifications.

Typical Performance Characteristics



Applications Information

LOW DRIFT CONSIDERATIONS

Achieving ultra-low drift in practical applications requires strict attention to board layout, thermocouple effects, and input guarding. For specific recommendations refer to AN-63 and AN-79.

A point worth stressing with regard to low drift specifications is testing of the LH0044. Simply stated—it is virtually impossible to test the device using a thermoprobe or other form of local heating. A one degree centigrade temperature gradient can account for tens of microvolts of virtual offset (or drift). The test circuit of *Figure 1* is recommended for use in a stabilized oven or continuously stirred oil bath with the entire circuit inside the oven or bath. Isothermal layout of the resistors is advised in order to minimize thermocouple induced EMF's.

OVER COMPENSATION

The LH0044 may be overcompensated in order to minimize noise bandwidth by paralleling the internal 100 pF capacitor with an external capacitor connected between pins 1 and 6. Unity gain frequency may be predicted by:

$$f = \frac{4 \times 10^{-5}}{100 \text{ pF} + C_{\text{ext}} \text{ pF}} \text{ (Hz)}$$

COMPENSATION

For closed loop gains in excess of 10, no external components are required for frequency stability. However, for gains of 10 or less, a 0.01 μF disc capacitor is recommended between pin 7 (V^+) and pin 8 (Comp). An improvement in ac PSRR will also be realized by use of the 0.01 μF capacitor.

OFFSET NULL

In general, further nulling of LH0044 is neither necessary nor recommended. For most applications the specified initial offset is sufficient.

However, for those applications requiring additional null, an obvious temptation might be to place a pot between pins 1 and 8 with the wiper returned to V^+ . This technique will usually result in reduced gain and increased offset drift due to mismatch in the TCR of the pot and R1 and R2. The technique is, therefore, not generally recommended.

The recommended technique for offset nulling the LH0044 is shown in *Figure 2*. Null is accomplished in A_2 and all errors are divided by the closed loop gain of the LH0044. Additional offset and drift incurred due to the use of A_2 is less than $1 \mu\text{V/V}$ for V^+ and V^- changes and $0.01 \mu\text{V}/^\circ\text{C}$ drift for the values shown in *Figure 2*.

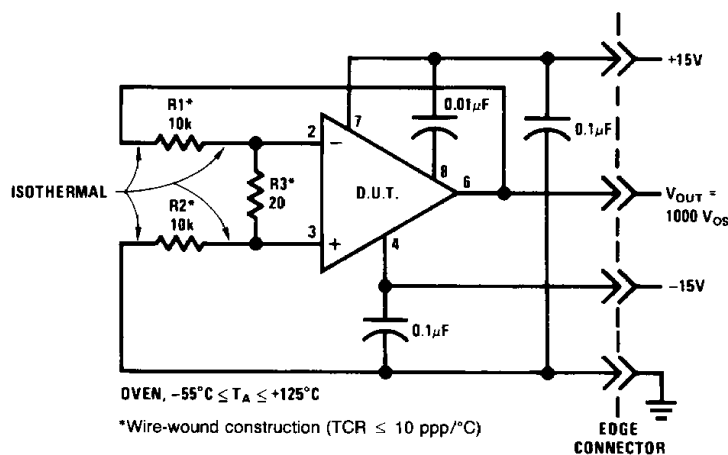


FIGURE 1. LH0044 Temperature Test Circuit

TL/K/5551-4

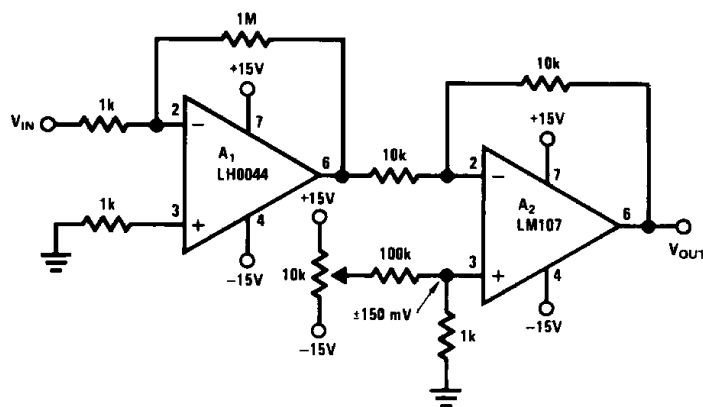
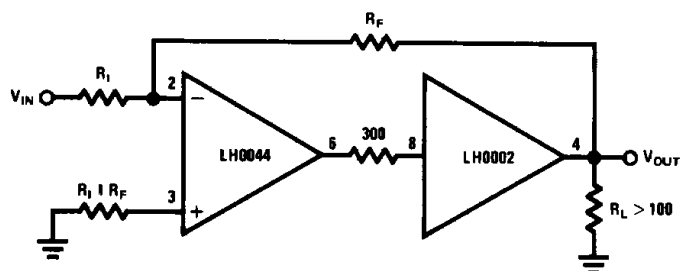


FIGURE 2. LH0044 Null Technique

TL/K/5551-5

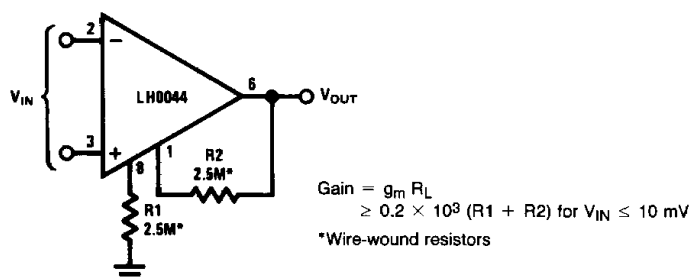
Typical Applications

Buffered Output for Heavy Loads



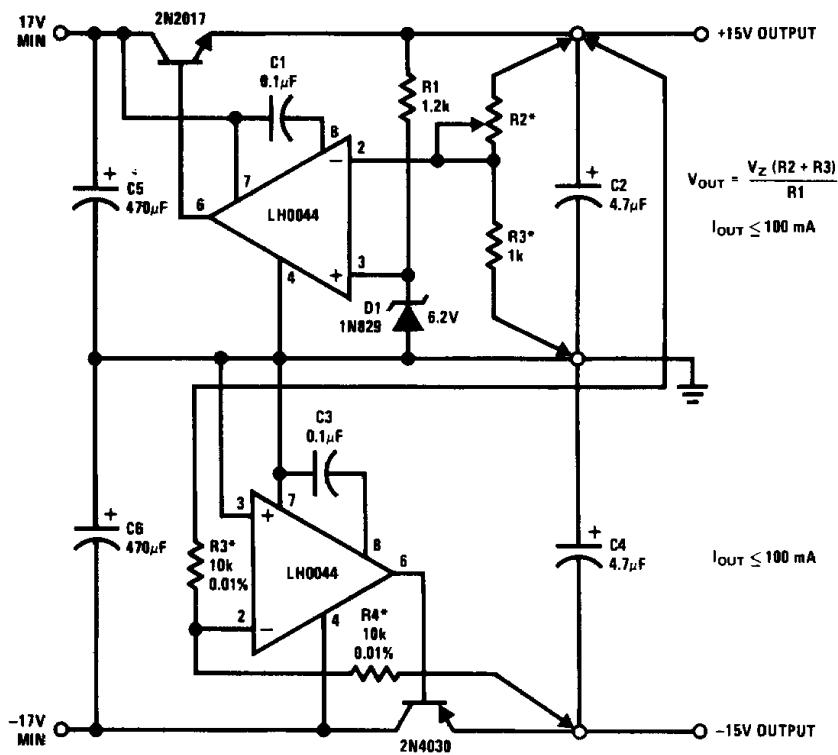
TL/K/5551-6

X1000 Instrumentation Amp



TL/K/5551-7

Precision Dual Tracking Regulator

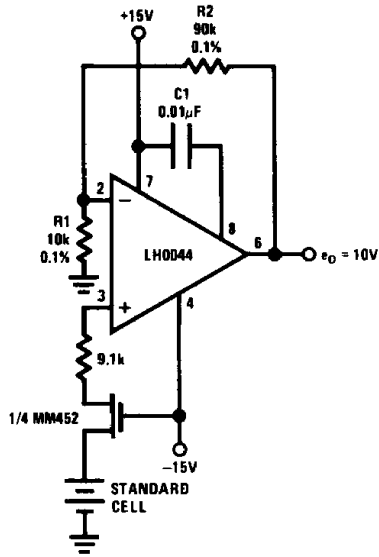


TL/K/5551-8

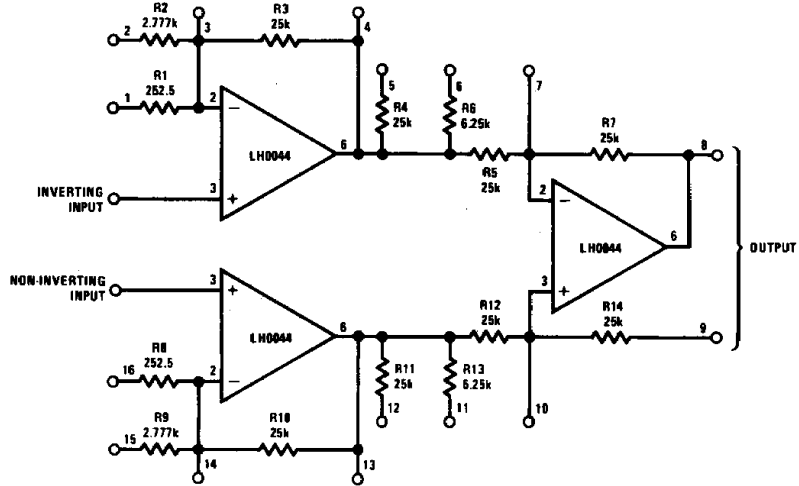
*Wire-wound for minimum drift.
 Line and load regulation $\leq 0.005\%$

Typical Applications (Continued)

10V Reference Supply



TL/K/5551-9

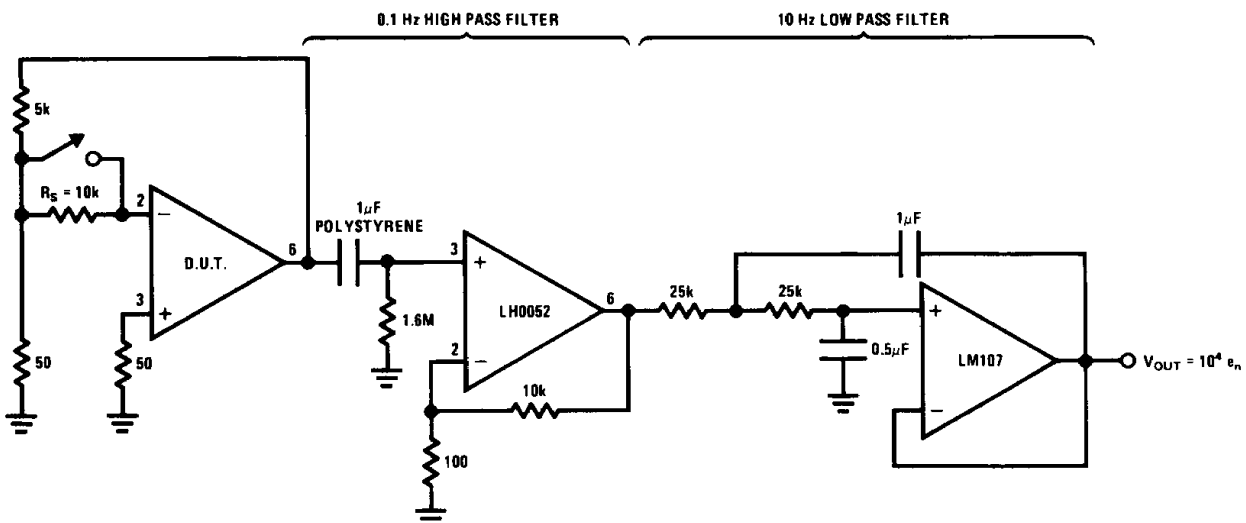


TL/K/5551-10

Precision Instrumentation Amplifier

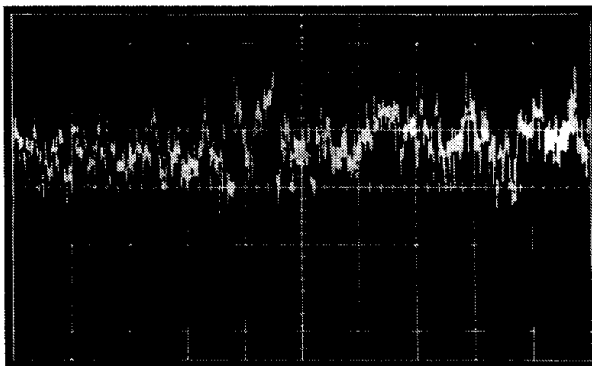
Overall Gain	Input Stage Gain	Output Stage Gain
X1	X1	X1
X2	X1	X2
X5	X1	X5
X10	X10	X1
X20	X10	X2
X50	X10	X5
X100	X100	X1
X200	X100	X2
X500	X100	X5
X995	X199	X5

Noise Test Circuit



TL/K/5551-11

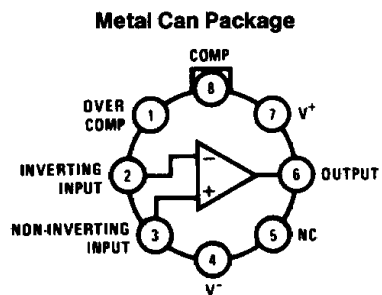
Noise Test Circuit (Continued)



VERT: 200 nV/DIV
HORIZ: 5 SEC/DIV

TL/K/5551-12

Connection Diagram



TL/K/5551-2

Top View

Case is electrically isolated

Note: Compensation is not normally required. However, for maximum stability, a $0.01\mu\text{F}$ capacitor should be placed between pins 7 and 8 when device is used below closed loop gains of 10.

**Order Number LH0044H, LH0044CH,
LH0044ACH or LH0044BH
See NS Package Number H08B**